

Comparative Analysis of Dynamic Performance Between Switched Reluctance Motors 6x4 and 8x6

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Abstract. This paper presents a comparative study between the Switched Reluctance Machine (SRM) 8x6 and 6x4 operating as a motor with both 60kW of power. The computer simulation was performed in Matlab™ / Simulink® environment and allowed the machine behavior under various operating conditions to be verified. This work made it possible to evaluate: phase voltage and current, speed, torque and efficiency. Comparative results between the two machines will be presented and discussed.

Key words. Electric vehicles, Energy efficiency, Switched Reluctance Motor 6x4, Switched Reluctance Motor 8x6.

1. Introduction

The growing demand for electricity from the lifestyle of modern society has made it necessary for more than one energy source to be studied and applied, taking into account the environmental and social impacts to be generated in the short and long term. As a result, the search for renewable energies has been rapidly expanding around the world such as solar, wind, biomass, geothermal and others[1]. In this context, several researches have been conducted in search of other alternatives such as the use of Switched Reluctance Motors in electric vehicles instead of the combustion motor.

In the last two decades, SRM have been on the rise due to their characteristics such as their structural simplicity, their robustness, phase-free tolerance, simplicity of control, low manufacturing cost, high temperature tolerance among others[2].

SRM construction is robust and inexpensive, assembled from a rotor and a stator, with protruding poles, non-oriented grain steel rolling and coils concentrated only on the stator poles. Its inductance is maximum when the rotor poles in relation to the stator are aligned and minimum when they are misaligned[3][4].

Due to its constructive characteristics, SRM has been increasingly explored in the area of electric traction. Several studies already perform their approach with the applications of electric traction, aiming to explore their particularities optimizing the machine according to the proposed application[5].

The machine is already being used in electric cars, city buses, trucks, backhoe loaders, railway locomotives, electric

trams, battery vehicles, and more [6][7][8][9]. Thus reducing emissions of polluting gases such as CO₂ that is harmful to human life by contributing to the greenhouse effect, which raises the temperature of the earth.

2. Switched Reluctance Machine

The SRM has a double overhang structure with rotor excitation using phase separated windings. Each phase of the machine contains its windings gathered only on the stator projections, while the rotor is composed only of ferromagnetic material, thus not containing coils or magnets [10]. Fig. 1 represents the structure of a 6x4 and 8x6 SRM.

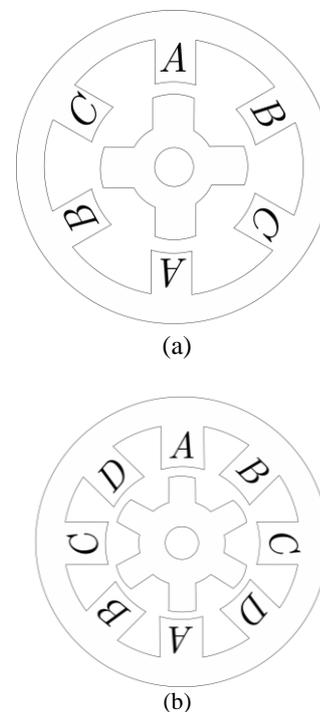


Fig.1. SRM 6x4 (a) and 8x6 (b) building structure.

In order to compare the two machines in question, the same parameters were considered. Table I shows the data of SRM 6x4 and 8x6 machines used in the present paper.

Table I. – Parameters of SRM 6x4 and 8x6.

Parameters	Values
Number of poles (stator and rotor)	8x6 e 6x4
Phase number	4 e 3
Output power	60KW
Stator resistance	0,05Ω
Moment of inertia	0,05 Kgm ²
Alignment inductance	23,6mH
Misalignment inductance	0,67mH

The reluctance of the SRM magnetic circuit is dependent on the rotor position relative to the stator. In the aligned position, when a pair of rotor poles are aligned with a pair of stator poles, the magnetic circuit has minimal reluctance and therefore maximum inductance. Fig.2 shows the characteristic inductance curves of the machine.

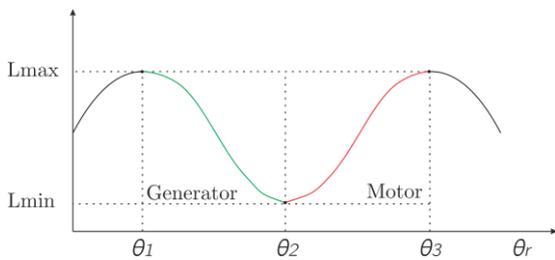


Fig.2. Inductance profile for an SRM.

3. Mathematical Model

For the development of the mathematical model was considered the two SRM operating as a motor. Fig.3 shows the equivalent phase circuit of the machine.

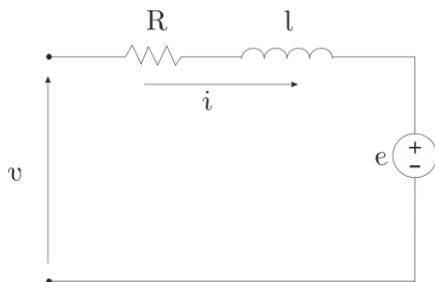


Fig.3. Equivalent circuit for one phase.

Initially, we can consider the voltage (v) of the SRM terminals as:

$$v = Ri + \frac{d\varphi}{dt} \tag{1}$$

where φ corresponds to the magnetic flux of the stator coils and R the stator winding resistance, i is the phase current.

We can define the magnetic flux as:

$$\varphi(i, \theta) = l(i, \theta)i(t) \tag{2}$$

Arranging the terms we can define the voltage of a phase being:

$$v = Ri + l(i, \theta) \frac{di}{dt} + e \tag{3}$$

$$e = i\omega \frac{dl(i, \theta)}{d\theta} \tag{4}$$

$$\omega = \frac{d(\theta)}{dt} \tag{5}$$

$$T_e = i^2 \frac{1}{2} \frac{dl(\theta)}{d\theta} \tag{6}$$

$$T_m = T_e - D\omega - J \frac{d\omega}{dt} \tag{7}$$

Where T_m is the mechanical torque, T_e is the electromagnetic torque, ω is the angular velocity, J is the moment of inertia and D is the viscous friction coefficient.

Using the equations of electrical and mechanical behavior of the Switched Reluctance Machine acting as motor. Reorganizing the equations and rewriting in the matrix form, equation 8, so that it facilitates the calculations for its solution [11].

$$\begin{bmatrix} v_a \\ v_b \\ v_c \\ v_d \\ T_m \\ 0 \end{bmatrix} = \begin{bmatrix} r_a & 0 & 0 & 0 & 0 & 0 \\ 0 & r_b & 0 & 0 & 0 & 0 \\ 0 & 0 & r_c & 0 & 0 & 0 \\ 0 & 0 & 0 & r_d & 0 & 0 \\ r_1 & r_2 & r_3 & r_4 & -D & 0 \\ 0 & 0 & 0 & 0 & -1 & 0 \end{bmatrix} \begin{bmatrix} i_a \\ i_b \\ i_c \\ i_d \\ \omega \\ \theta \end{bmatrix} + \tag{8}$$

$$\begin{bmatrix} l_a(i, \theta) & 0 & 0 & 0 & 0 & i_a \frac{\partial l_a(i, \theta)}{\partial \theta} \\ 0 & l_b(i, \theta) & 0 & 0 & 0 & i_b \frac{\partial l_b(i, \theta)}{\partial \theta} \\ 0 & 0 & l_c(i, \theta) & 0 & 0 & i_c \frac{\partial l_c(i, \theta)}{\partial \theta} \\ 0 & 0 & 0 & l_d(i, \theta) & 0 & i_d \frac{\partial l_d(i, \theta)}{\partial \theta} \\ 0 & 0 & 0 & 0 & -J & 0 \\ 0 & 0 & 0 & 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} \dot{i}_a \\ \dot{i}_b \\ \dot{i}_c \\ \dot{i}_d \\ \dot{\omega} \\ \dot{\theta} \end{bmatrix} \tag{8}$$

$$r_1 = \frac{1}{2} \cdot i_a \frac{\partial l_a(i, \theta)}{\partial \theta}, r_2 = \frac{1}{2} \cdot i_b \frac{\partial l_b(i, \theta)}{\partial \theta},$$

$$r_3 = \frac{1}{2} \cdot i_c \frac{\partial l_c(i, \theta)}{\partial \theta}, r_4 = \frac{1}{2} \cdot i_d \frac{\partial l_d(i, \theta)}{\partial \theta} \tag{9}$$

To optimize the machine performance, current control was used in only one of the coils of each phase, the others being driven directly by the position sensor signal. The diagram of this block is shown in Fig.10.

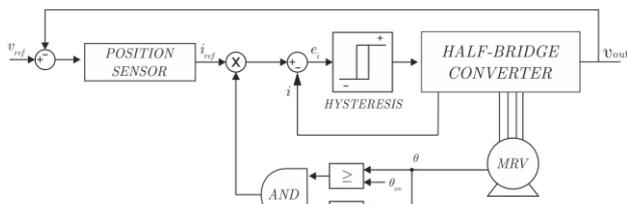


Fig.10. Current Controller Diagram.

C. Speed control

To control the speed of the SRM was used the proportional and integral controller (PI), which causes the system to reach the reference speed specified by the operator. The control acts on the power of the machine causing the speed to change depending on the excitation of the machine. Fig.11 shows the system loop diagram with the PI control.

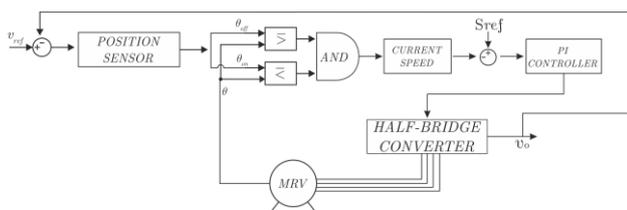


Fig.11. PI controller diagram for speed.

D. Efficiency

To estimate the performance of the SRM a block was developed to calculate the efficiency (η) of each machine, represented by equation 11, where the input power is calculated by the product of the converter current and supply voltage, already output power is the product of motor speed with torque.

$$\eta = \frac{T_e * \omega_r}{v * i} \quad (11)$$

E. Switching System

The most common SRM drive is the system known as single pulse drive. In this system, during inductance growth, both switches of the HB converter. receive at their control terminals a pulse to allow the phase coil to magnetize or demagnetize. The need to control some quantities such as current speed, and SRM conjugate requires that some switching techniques be employed to perform the control. To avoid the machine is to perform the control and it is necessary to switch keys, and this ends up generating losses in the process. Therefore it is necessary to look for techniques that minimize switching losses.

Thus, the soft chopping switching system was employed, in which the lower switch of the HB converter is kept constant during the driving angle, while the upper switch of the converter is switching according to the required control need. The system can be observed in the Fig.12.

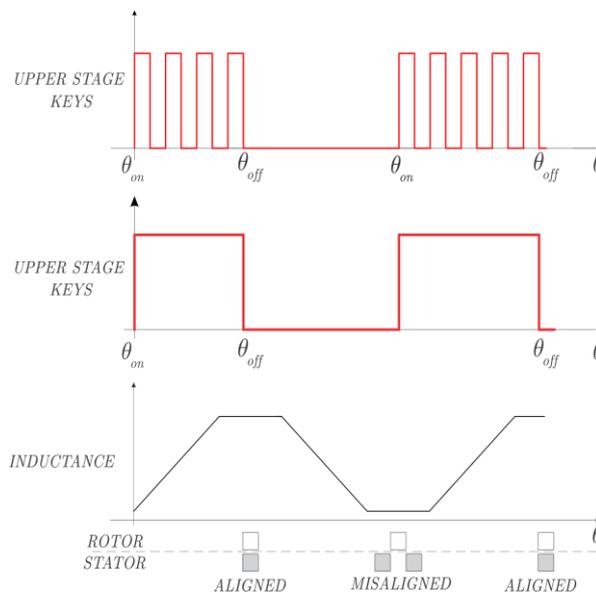


Fig.12. Switching system waveforms.

In general, this method presents lower levels of acoustic noise and electromagnetic interference. Such a switching strategy also has lower ripple levels in the current and affects the performance of the machine [12][13].

5. Simulation Results

Next, we will show several curves that validate the experimental simulation. Initially, Fig.13 shows the action of the PI control on two machines with equal operating conditions, where the speed setpoint remains constant at 500rpm until 0.3 seconds, and the setpoint is changed to 2000 rpm and remains at this speed. Speed up to 0.6 seconds in sequence and set the speed to 1600 rpm and keep it that way.

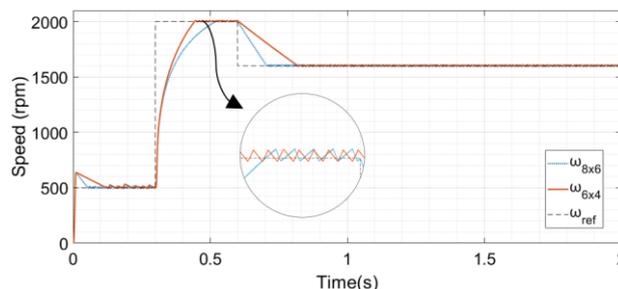


Fig.13. SRM speed with PI controller actuation.

The performance of the machines is shown in Fig.14. The curves were plotted with the same operating conditions used for speed control.

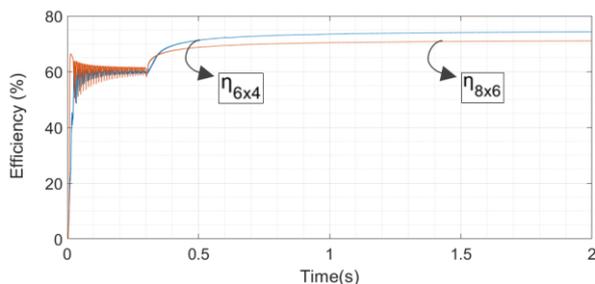


Fig.14. SRM efficiency.

The Fig. 15, represents the efficiency of the SRM 8x6 applied to the new drive system, it is possible to verify that using the soft chopping system there was an increase in its yield, caused by the reduction of losses in the system.

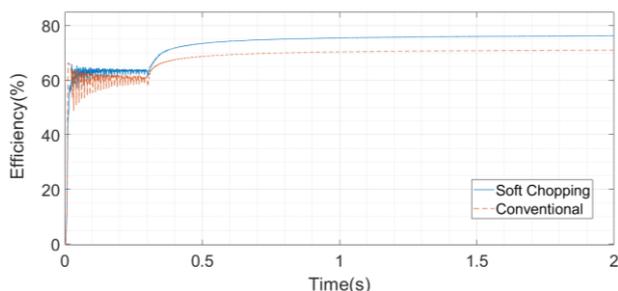


Fig.15. SRM 8x6 efficiency with soft chopping

4. Conclusion

This paper presents a comparative study between two 60 kW SRM, one of them is a three-phase 6x4 pole machine and the other an 8x6 four-phase machine.

The computer simulation was modeled from the magnetization curves of the machines. Graphs were generated using current and speed control.

Analyzing the results obtained through the computer simulation it was verified that the SRM 8x6, presents less oscillation in the conjugate than the SRM 6x4, under the same operating conditions. In addition, it was found that SRM 8x6 has lower performance than SRM 6x4.

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